



synchro/resolver amplifier
5VA

series 175A600

FEATURES

- Synchro amplification/voltage conversion
- Resolver amplification/voltage conversion
- Synchro/resolver and resolver/synchro conversion
- Full 5VA Output
- Transformer Isolated Output
- "Locked Rotor" Protection
- Short Circuit and Overload Protection
- Thermal Cutoff Protection

APPLICATIONS

- Training Simulators
- Fire Control Systems
- Retransmission Systems
- Remote Indicators

GENERAL DATA

The series 175A600 is a 2.62" x 3.12" x 0.8" module designed for p.c. board mounting. The amplifier employs high balanced input impedances and low balance output impedances that minimize loading errors. It is designed to deliver a full 5VA making it capable of driving virtually all 400Hz size 8 and 11 control transformer (CT), control differential transmitter (CDX), torque receiver (TR) synchro loads, resolver control transformer (RC) and resolver differential transmitter (RD) loads while maintaining a ± 4 arc minute accuracy. The synchro/resolver amplifier is current limited and features a thermal cut-off to prevent overheating. An aluminum top plate on the module provides all the necessary heat-sinking.

THERMAL CONSIDERATIONS

The top of the 175A600 consists of a aluminum plate providing all the required heat sinking. Thermal resistance top plate to free air is 15°C/VA. A thermal cutout is incorporated that disables the output power amplifiers when the top plate reaches 105°C. The output is automatically restored when the temperature drops below 100°C. The aluminum top plate should be provided with sufficient air circulation. The thermal resistance of the top plate may be decreased by a factor of three or greater by simply blowing air of sufficient velocity over the top plate.

INPUT/OUTPUT CONFIGURATIONS

The amplifier can be supplied as any one of four of the following configurations:

RESOLVER INPUT RESOLVER OUTPUT

$$\begin{aligned} \text{SIN/COS} \quad E_{S1-S3} &= KE(\text{SIN})_{\ominus} \\ E_{S4-S2} &= KE(\text{COS})_{\ominus} \end{aligned}$$

RESOLVER INPUT SYNCHRO OUTPUT

$$\begin{aligned} \text{SIN/COS} \quad E_{S1-S3} &= KE(\text{SIN})_{\ominus} \\ E_{S2-S3} &= KE(\text{SIN})_{(\ominus+120^{\circ})} \\ E_{S1-S2} &= KE(\text{SIN})_{(\ominus+240^{\circ})} \end{aligned}$$

SYNCHRO INPUT RESOLVER OUTPUT

$$\begin{aligned} \text{S1-S2-S3} \quad E_{S1-S3} &= KE(\text{SIN})_{\ominus} \\ E_{S4-S2} &= KE(\text{COS})_{\ominus} \end{aligned}$$

SYNCHRO INPUT SYNCHRO OUTPUT

$$\begin{aligned} \text{S1-S2-S3} \quad E_{S1-S3} &= KE(\text{SIN})_{\ominus} \\ E_{S2-S3} &= KE(\text{SIN})_{(\ominus+120^{\circ})} \\ E_{S1-S2} &= KE(\text{SIN})_{(\ominus+240^{\circ})} \end{aligned}$$

Where: K = output voltage/input voltage
E = input voltage
 \ominus = synchro/resolver shaft angle

SPECIFICATIONS

| Parameter | Value |
|--------------------------------------|--|
| Accuracy¹ | |
| Passive (CT CDX RC RD) | ±4 arc minutes |
| Active (TR) | ±12 arc minutes |
| Analog Inputs | |
| Synchro (S1-S2-S3) ² | 11.8Vrms L-L 400Hz @ 100KΩ 90.0Vrms L-L 400Hz @ 600KΩ |
| Resolver (SIN/COS) ³ | 11.8Vrms 400Hz @ 100KΩ |
| Analog Outputs⁴ | |
| Synchro (S1-S2-S3) | 11.8Vrms L-L 400Hz @ 3Ω 90.0Vrms L-L 400Hz @ 130Ω |
| Resolver (S1-S2-S3-S4) | 11.8Vrms 400Hz @ 3Ω |
| Output Loading | |
| 11.8V Output | |
| CT CDX RC RD | 28Ω min. (Z _{so}) |
| TR | 3Ω min. (Z _{ss}) |
| 90V Output | |
| CT CDX RC RD | 1620Ω min (Z _{so}) |
| TR | 130Ω min. (Z _{ss}) |
| Power Supply Req_t. | |
| Voltage | ±15Vdc ±5% |
| Current | 60mA + load |
| Peak Current | 2A |
| Temperature Ranges | |
| Operating | 0° to 70°C |
| Storage | -55° to +125°C |

NOTES:

- Accuracy applies for:
 - ±5% power supply voltage variation.
 - ±10% signal amplitude variation.
 - ±10% frequency variation.
 - 10% signal harmonic distortion.
 - any balance load no load to full load.
 - over operating temperature range.
- Synchro input is solid state Scott T, common mode voltages up to specified L-L voltage have no effect on operation.
- Resolver input is solid state differential, SIN/COS returns are to be connected to the COM pin. Do not connect COM to GND
- Synchro and resolver outputs are transformer isolated. Transformation ratio accuracy is ±2%.

POWER SUPPLIES

The short circuit current of the 175A600 is 2A maximum. The ±15Vdc power supplies should be capable of 2A peaks. On power-up instantaneous 2A peaks can occur. The absence of any one of the supply voltages will not cause damage or excessive current flow from the other supply.

DRIVING TR LOADS

The amplifier is designed to drive synchro torque receivers (TR). The 175A600 can drive any 11.8V 400Hz TR whose Z_{ss} is greater than 3Ω or any 90V 400Hz TR whose Z_{ss} is greater than 130Ω. The value of Z_{ss} can be obtained from the synchro manufacturer's specification data sheet. If Z_{ss} is not specified it can be approximated from the specified rotor and stator resistances as follows:

$$Z_{ss} (11.8V) = \text{stator resistance} + (0.21)\text{rotor resistance}$$

$$Z_{ss} (90.0V) = \text{stator resistance} + (0.61)\text{rotor resistance}$$

In the null condition, when the 175A600 output angle equals the TR shaft angle theoretically no circulating currents will flow in the stator windings. However, at power-up large initial angular offsets between the amplifier output and the TR rotor shaft angle can occur. If this offset angle is too large the amplifier can't deliver enough current causing a hang-up condition known as "locked rotor" where the TR rotor shaft is not driven back to null but stalls at some other angle. The 175A600 prevents the possibility of this hang-up condition by employing a kick circuit that momentarily shifts its output by 90° when a sustained over-current condition exists. Without the kick circuit, larger drive currents would be required, therefore the kick circuit is an important asset in reducing the size of the synchro amplifier.

To determine the power required for an off null condition the following equation can be used:

$$VA = \frac{0.866(V_{LL})^2 \sin \frac{\Theta}{2}}{Z_{SS} + Z_{OUT}}$$

Where: V_{LL} = amplifier analog output voltage

Θ = offset angle

Z_{SS} = stator impedance, rotor shorted

Z_{OUT} = amplifier output impedance

Caution: Both synchro amplifier analog input signals and torque receiver reference excitation must be present otherwise large circulating currents will flow in the stator windings.

DRIVING CT AND CDX LOADS

When driving CT and CDX loads the 175A600 must have enough steady state power capability to drive the Z_{SO} of the device. Generally a CT will be lightly loaded and the following equation can be used to calculate its VA requirement:

$$VA = \frac{0.866(E_{LL})^2}{Z_{SO}}$$

Where: E_{LL} = CT stator (primary) input voltage
 Z_{SO} = CT stator input impedance

CT's are highly inductive loads and it is possible to save power by tuning such loads. Three capacitors may be placed across the legs of the CT in a delta configuration. The value of the capacitance is given by:

$$C = \frac{X_L}{4\pi f [R^2 + (X_L)^2]}$$

Where: C = capacitance in Farads
 f = frequency in Hz
 R = real component of Z_{SO}
 X_L = reactive component of Z_{SO}

Relatively stable capacitors should be used and must be capable of withstanding the full AC voltage.

When the load has been tuned more loads can be driven in parallel because the load impedance is increased to:

$$Z_{SO} = \frac{R^2 + (X_L)^2}{R}$$

The output of a CDX is often loaded with another synchro. Therefore when computing the VA requirement, both the CDX and its load impedance must be taken into account as follows:

$$VA = \frac{0.866(E_{LL})^2(Z_{SO} + Z_{SO'})}{Z_{SO} \times Z_{SO'}}$$

Where: E_{LL} = CDX stator (primary) input voltage
 Z_{SO} = CDX stator input impedance
 $Z_{SO'}$ = CDX load impedance

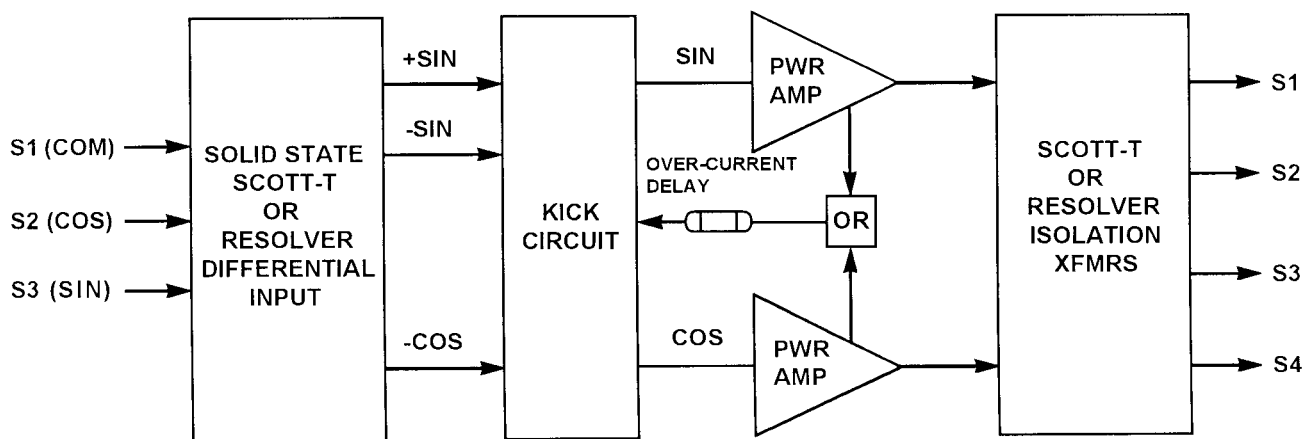
DRIVING RC AND RD LOADS

When driving resolver control transformers (RC) or resolver differentials (RD) the 175A600 must have enough steady state power capability to drive the input impedance (Z_{SO}) of the device. The following equation can be used to calculate their VA requirements:

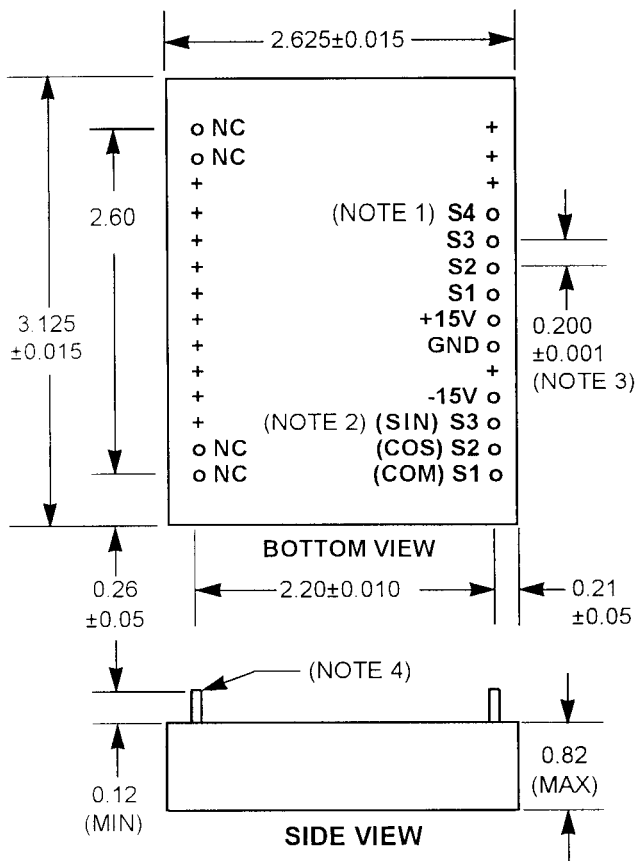
$$VA = \frac{1.41(E_{LL})^2}{Z_{SO}}$$

Where: E_{LL} = stator input (primary) input voltage
 Z_{SO} = stator input impedance

BLOCK DIAGRAM



MECHANICAL OUTLINE



NOTES:

- (1) Synchro/resolver output pins. S4 pin appears on resolver output models only.
- (2) Synchro/resolver input pins. S1-S2-S3 designators are for synchro input models. SIN, COS and COM designators are for resolver input models.
- (3) Non-cumulative.
- (4) Rigid 0.040 diameter pins suitable for solder-in or plug-in applications.
- (5) Dimensions are in inches.
- (6) NC pins have no internal contact.
- (7) Weight = 7.5oz.

ORDERING INFORMATION

| 175A SUFFIX | INPUT | | OUTPUT | | FREQ (HZ) |
|----------------|-------|-------|--------|-------|--|
| | TYPE | VOLTS | TYPE | VOLTS | |
| 600 | SYN | 11.8V | SYN | 11.8V | 400HZ ↑ ↓ 400HZ |
| 601 | SYN | 11.8V | SYN | 90.0V | |
| 602 | SYN | 90.0V | SYN | 11.8V | |
| 603 | SYN | 90.0V | SYN | 90.0V | |
| 604 | SYN | 11.8V | RES | 11.8V | |
| 605 | SYN | 90.0V | RES | 11.8V | |
| 606 | RES | 11.8V | RES | 11.8V | |
| 607 | RES | 11.8V | SYN | 11.8V | |
| 608 | RES | 11.8V | SYN | 90.0V | |
| 609 | RES | 6.0V | SYN | 11.8V | |
| 610 | RES | 6.8V | SYN | 11.8V | |
| 611 | RES | 6.0V | SYN | 90.0V | |
| 612 | RES | 6.8V | SYN | 90.0V | |

NOTES:

- (1) Standard temperature range is 0° to $+70^{\circ}\text{C}$, add suffix IT to part number for (-25° to $+85^{\circ}\text{C}$) temp. range.
- (2) Other input/output voltages available on special order, consult factory.

WARRANTY

All units warranted against defects in materials and workmanship for 1 year from the date of shipment. Liability is expressly limited to servicing, adjusting or replacing any CSI product returned to our factory with delivery charges prepaid. In no case shall our liability exceed the original price.